

Ultracompact, EEG-compatible NIRS System



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OBJECTIVE

The development of a miniaturized, portable NIRS system, to be integrated with EEG for BCI applications. Here, we present our most recent step toward this goal, a table-top multi-channel NIRS imager utilizing direct LED illumination and digital signal processing.

INTRODUCTION

Principles of Near-Infrared Spectroscopy (NIRS):

- Uses low-energy optical radiation (~700-900 nm)
- Scatter-dominated light propagation in tissue
- Transmission up to few cm
- Sensitive to absorption changes caused by hemodynamics: measures relative changes in oxygenated and de-oxygenated Hemoglobin (ΔHbO , ΔHbR)
- Requires contact based measurements through 'optodes' (usually fiber optic bundles, sometimes integrated electronic sensors)
- One optode pair (transmitter (source) + receiver (detector)) constitutes one data channel.

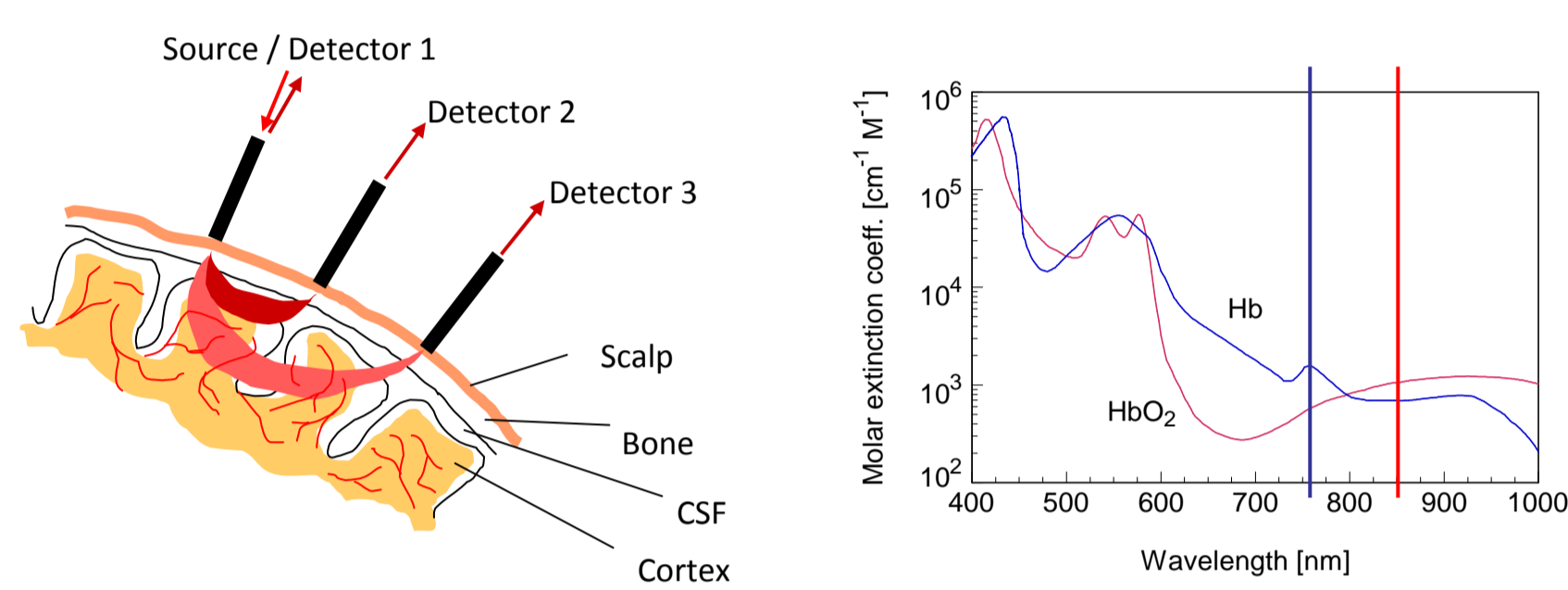


Figure 1. Left: Schematic depiction of NIRS imaging of brain hemodynamics. Right: Spectra of HbO and HbR; vertical lines indicate imaging wavelengths.

NIRS Promises:

- Non-invasive, harmless
- Good time resolution (~10 Hz)
- Can be made small, inexpensive
- Little subject restriction, long term monitoring
- Integrates well with other methods (EEG, MRI,...)

NIRS Challenges:

- Penetration depth ~cm
- Spatial resolution ~mm-cm
- Sensitive to artifacts from motion, surface- near 'global' hemodynamics
- Probe setup

INSTRUMENTATION

- Size: 33 cm (L) × 27 cm (W) × 17 cm (H)
- Mass < 10 kg
- Channels: 1-8 sources, 1-16 detectors
- Scan rate: > 8 Hz (8 sources)
- Illumination: 760/850 nm LED
- Digital signal demodulation



Figure 2. Picture of the instrument. Shown is version with 2 sources, 4 detectors, and 4 trigger connectors for experiment synchronization.

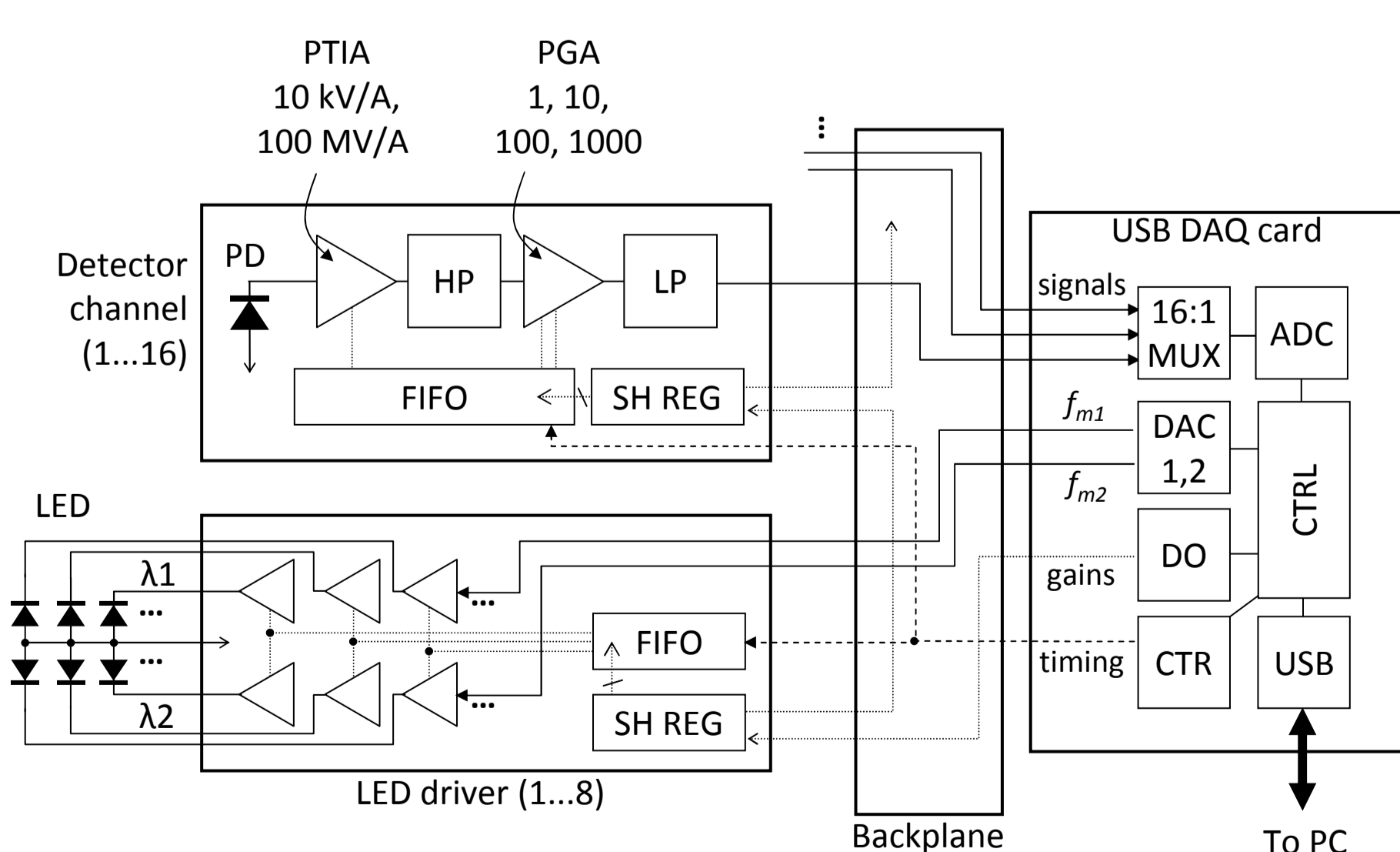


Figure 3. Functional diagram of NIRS imager (omitted: power supply, trigger channels). PD: photo diode; PTIA: programmable transimpedance amplifier; PGA: programmable gain amplifier; HP: RC high-pass filter ($f_{3dB} = 482$ Hz); LP: 4th order Chebyshev low-pass filter ($f_{pass} = 2.0$ kHz); FIFO: first-in-first-out buffer; SHREG: serial-to-parallel shift register; f_{mod} : modulation frequencies; MUX: multiplexer; ADC: analog-to-digital converter; DAC: digital-to-analog converter; DO: digital out lines; CTR: counter; CTRL: controller.

Illumination

- Time-multiplexed source position encoding
 - Simultaneous, frequency-encoded dual-wavelength illumination (760, 850 nm):
- $$P_{i,j} = P_{dc} + P_{mod} \sin(2\pi f_m / f_s j + \varphi_i)$$
- Direct tissue contact light emitting diodes (LED)
 - 90° Optode head design integrates into EEG electrode
 - LEDs vs. Lasers:
 - ~10 mW total optical power per λ @ ~50 mA
 - Spectral width ~30 nm
 - Emission angle (FWHM) $\pm 20^\circ$

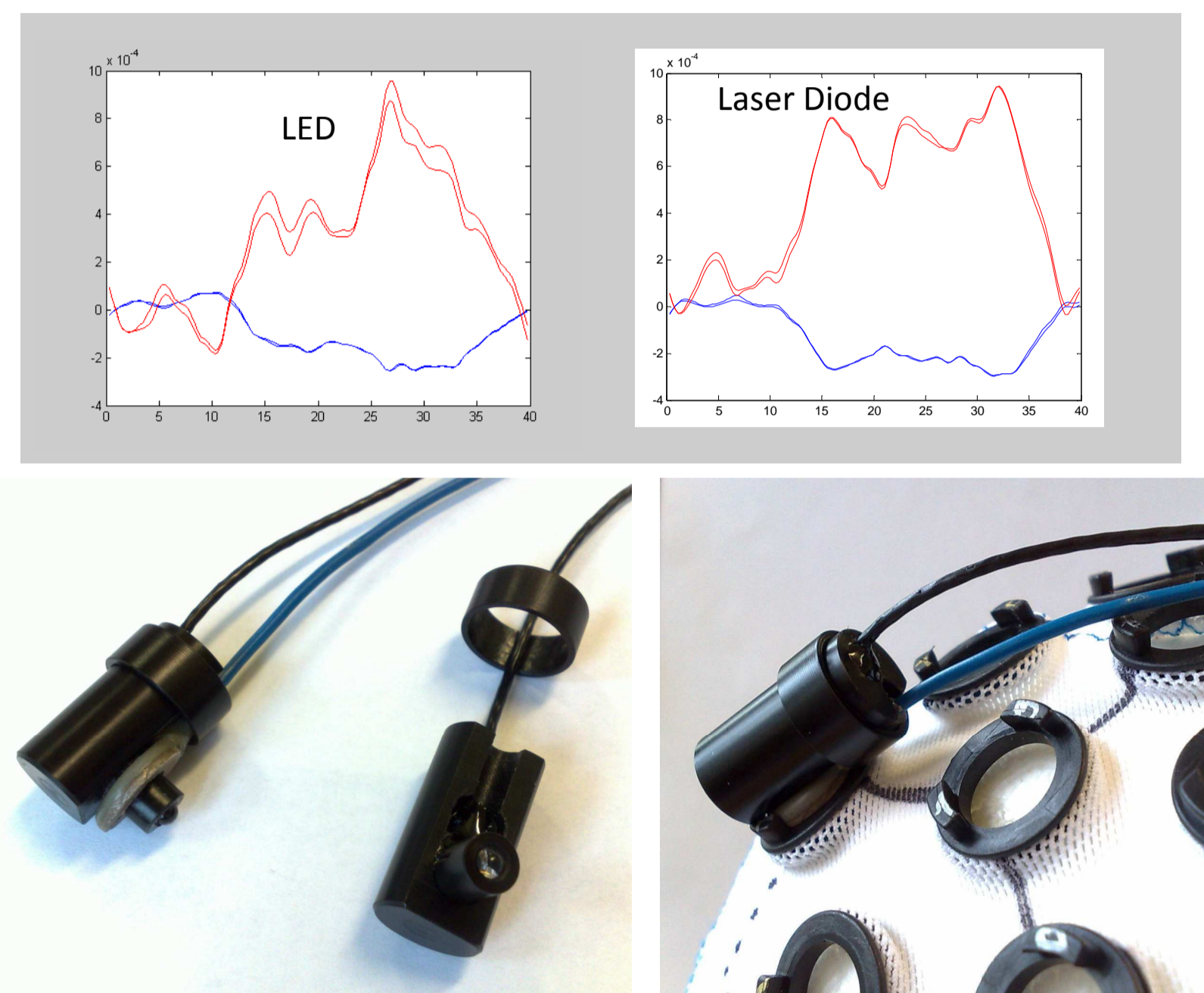


Figure 4. Top row: Hand gripping task result illustrating comparable crosstalk and separability for LED (760/850 nm, left) vs. Laser (760/830 nm, right) illumination. Bottom row: Low-profile LED optodes, from left to right: Connected to ring electrode; optode only with retaining ring; mounted to electrode in EEG cap.

Detection and Signal Processing

- Fiber-optic bundles ($\varnothing = 2.4$ mm) with bent tip
- Photo diode in unbiased photovoltaic mode (large dynamic range)
- Gain-switched amplification maximizes dynamic range
- Direct analog-to-digital conversion of modulated signal
- Demodulation in PC software (LabVIEW™) following an algorithm described in [Lasker et al.]:

1. Component-wise multiplication of signal vector \mathbf{V} and synthesized orthogonal reference signals

$$I_{i,j} = \cos(2\pi f_m / f_s j) \text{ and } Q_{i,j} = \sin(2\pi f_m / f_s j):$$

$$X'_{i,j} = V_j I_{i,j}; \quad Y'_{i,j} = V_j Q_{i,j}$$

2. Time-average the resulting vectors to obtain the in-phase and quadrature amplitudes

$$X_i = 1/N \sum_j^N X'_{i,j}; \quad Y_i = 1/N \sum_j^N Y'_{i,j}$$

3. Compute the magnitude to obtain the detector reading $R_{d,i}$:

$$R_{d,i} = \sqrt{X_i^2 + Y_i^2}$$

4. Averaging *exactly* over an interval that is an integer multiple of the modulation period

$$N = k / f_{m,i} \quad (\text{where } k = 1, 2, \dots)$$

and using commensurable modulation frequencies results in *perfect* wavelength discrimination.

Noise considerations:

- Detection limit is determined by electronic noise at highest gain
 - Dominant noise source is thermal (Johnson) noise of the feedback resistor. Detection bandwidth BW is determined by low-pass RC filter ($f_{3dB} = 480$ Hz) and anti-aliasing filter (4th order Chebyshev $f_{cutoff} = 2.0$ kHz):
- $$BW = 2.0 \text{ kHz} - 2/\pi \cdot 0.48 \text{ kHz} = 1.25 \text{ kHz}$$
- $V_n = \sqrt{4k_B T R BW}$. For $T=300\text{K}$, $R=10^8 \Omega$: $V_n = 46 \mu\text{V}_{rms}$
 - Highest gain = 1000 $\rightarrow V_{n,gain} = 46 \text{ mV}_{rms}$
 - Averaging over 100 samples reduces noise by factor 10 $\rightarrow V_{n,demod} = 4.6 \text{ mV}_{rms}$ (Measured: $V_n = 5 \text{ mV}_{rms}$)
 - Noise equivalent power (NEP): $V_{sig} = V_{noise} = 5 \text{ mV}_{rms}$
 $\rightarrow V_p = \sqrt{2} V_{rms} = 7 \text{ mV}_p \rightarrow I_p = V_p / 10^{11} \Omega = 7 \times 10^{-14} \text{ A} \rightarrow \text{NEP} = 2\text{W/A} \times I_p = 0.14 \text{ pW}_p$

RESULTS

Instrument Performance

- Sensitivity: better 0,5 pW NEP
- Stability: better 0.01%
- Drift: < 1%/hr after warm up
- Dynamic range (theo): $[1 \text{ V} / 10^4 \text{ V/A}] / [5 \text{ mV} / 10^{11} \text{ V/A}] = 2 \times 10^9 = 93 \text{ dB}_{pwr}$

Simultaneous NIRS-EEG validation

Paradigm:

- Alternating self-paced, visually cued finger tapping
- Left (20 s) \rightarrow rest (20 s) \rightarrow right (20 s) \rightarrow rest (20 s)... [9 times]

NIRS measurement:

- 2 Sources, 4 Detectors (per hemisphere: 1 S x 2 D), $f_{sample} = 22.5$ Hz
- Customized EEG cap (EASYPAC GmbH) for electrode/optode placement (Fig. 5)
- Signal analysis in MATLAB™ based NILAB (Charite Berlin): Band pass filter, modified Beer-Lambert law, block-averaging

EEG measurement:

- 13-channel recording with BrainAmp (Brain Products GmbH): $f_{sample} = 1.0$ kHz, FCz as reference, BW = DC to 250 Hz
- Band pass filter [0.5 Hz, 100Hz], epoched for each condition (-3 s to +25 s relative to stimulus onset)
- Wavelet-based time-frequency (TF) analysis (Morlet, 12 cycles, 5-25 Hz) on single trial basis
- Averaging of single trial TF results; baseline (-1 to 0 s) subtraction

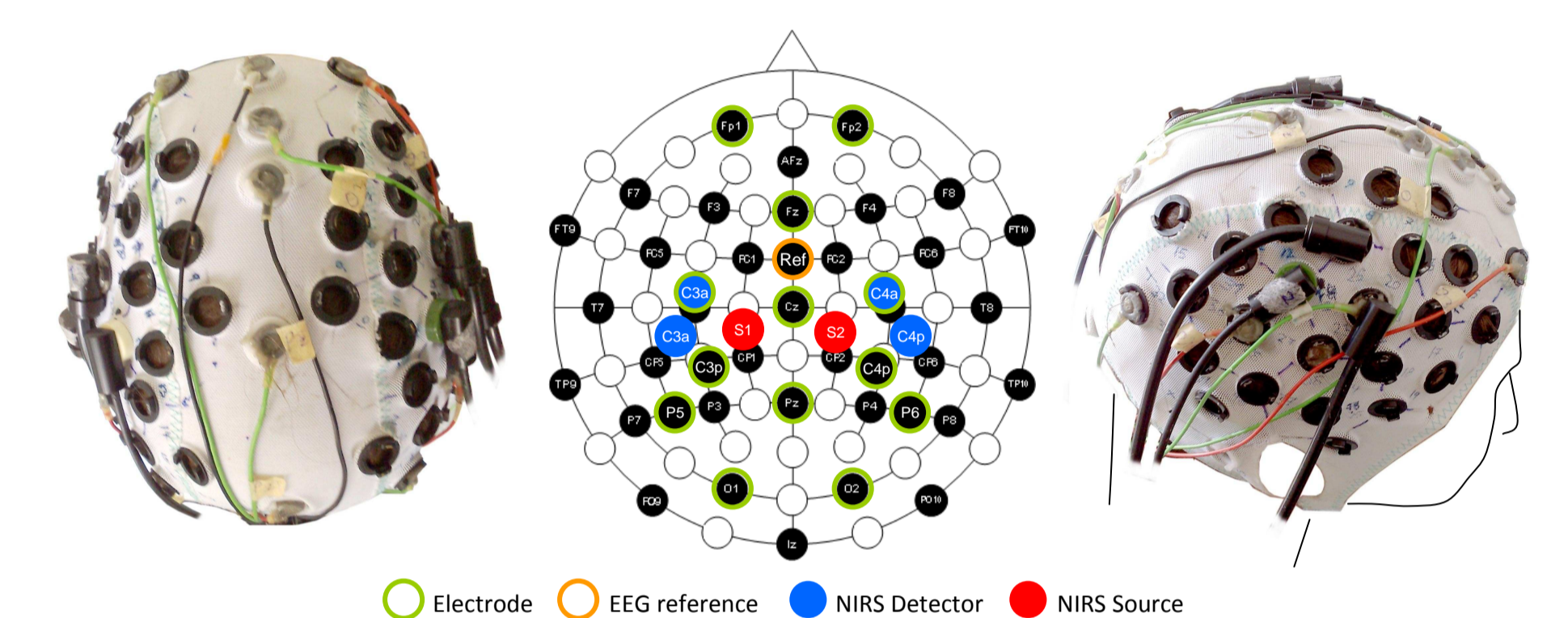


Figure 5. Electrode and optode placement on EEG cap.

Experimental Results

EEG

- No spectral interferences from LED optode
- Sustained μ -desynch (α - & β -range) during contralateral tap

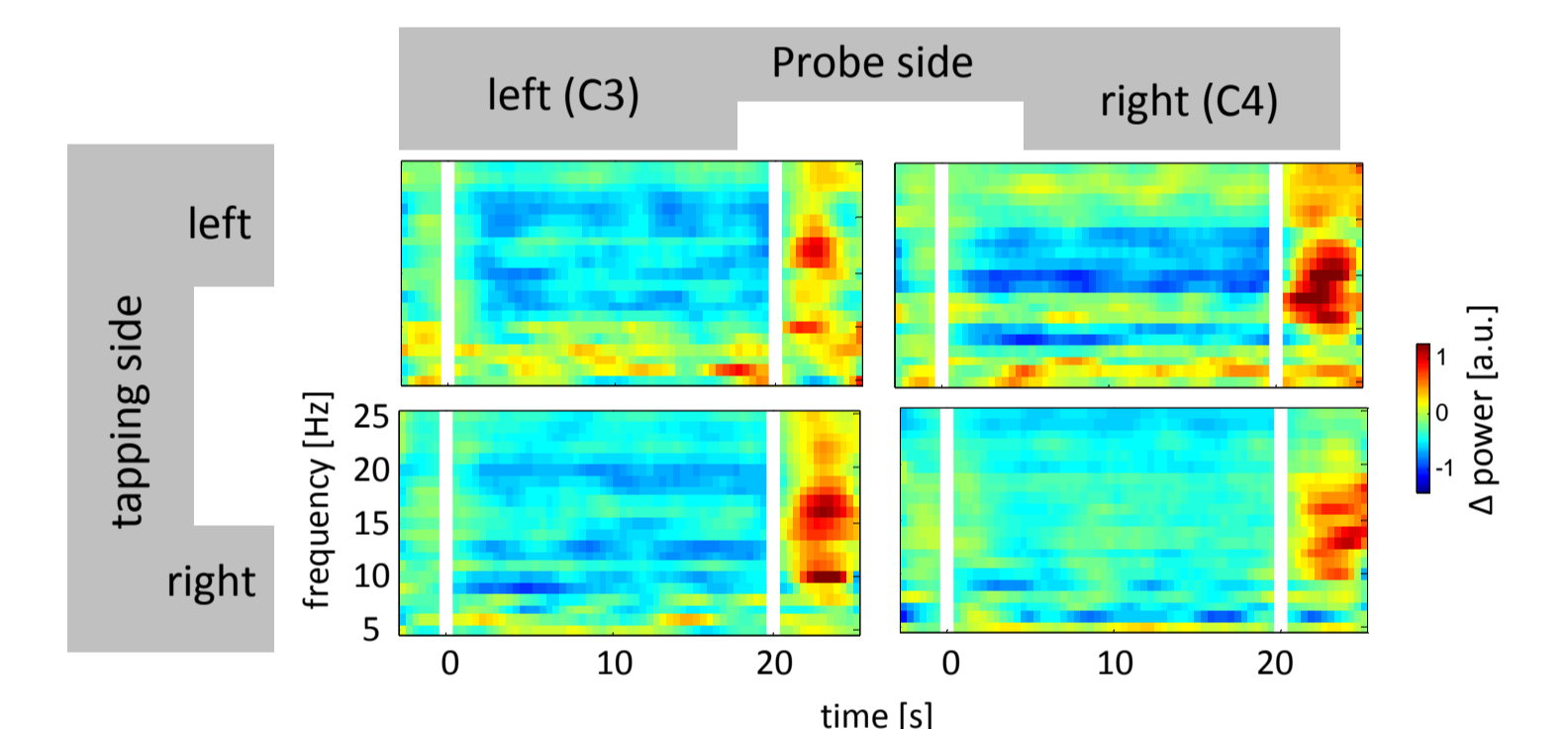


Figure 6. EEG time-frequency results.

NIRS

- Prototypical activation for motor task is observed: $\text{HbO} \uparrow$ & $\text{HbR} \downarrow$; sustained response to tapping
- HbO : Susceptible to physiological noise, therefore less indicative for tapping side
- HbR : Stronger decrease contralateral to tapping side
- Robust, condition related single-trial response

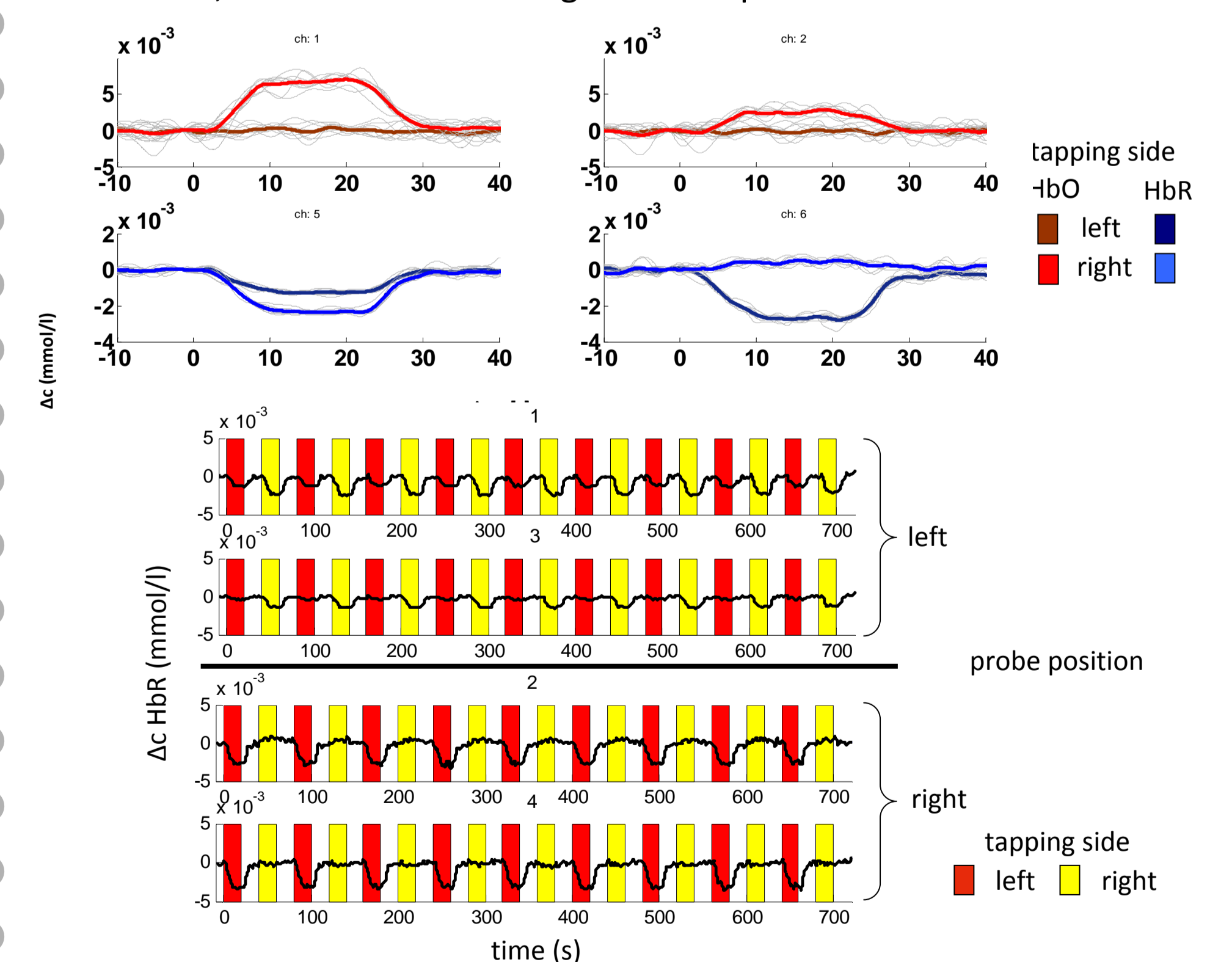


Figure 7. NIRS results. Top: Block-averaged vs. single-trial response. Bottom: Continuous time courses for nine consecutive trials.

DISCUSSION

- No interferences in EEG signals
- Excellent signal-to noise ratio in HbO and HbR
- Fast and flexible probe placement for concurrent EEG/NIRS measurement
- Robustness of single trial response promising for real-time applications (BCI, neuro-feedback)
- Portability/compactness allows field studies